DIRECT TORQUE CONTROL OF MATRIX CONVERTER FED INDUCTION MOTOR DRIVE

Suresh M.¹, Karpagam.J², Nirmal Kumar.A.³

¹,²Department of EEE, Bannari Amman Institute of Technology, Sathyamangalam, TamilNadu.
³Department of EEE, Info Institute of Engineering, Coimbatore, TamilNadu.

Abstract

This paper presents the direct torque control of induction motor drive using matrix converters. DTC is a high performance motor control scheme with fast torque and flux responses. However, the main disadvantage of conventional DTC is electromagnetic torque ripple. In this paper, direct torque control for Induction Motors using Matrix Converters is analysed and points out the problem of the electromagnetic torque ripple which is one of the most important drawbacks of the Direct Torque Control. Besides, the matrix converter is a single-stage ac-ac power conversion device without dc-link energy storage elements. Matrix converter (MC) may become a good alternative to voltage-source inverter (VSI). This work combines the advantages of the matrix converter with those of the DTC technique, generating the required voltage vectors under unity input power factor operation. Simulation results demonstrates the effectiveness of the torque control.

Keywords: Direct Torque control, vector control, Matrix Converters, space vector modulation.

I. INTRODUCTION

For many years, induction machine[6,8] have provided the most common form of electromechanical drive for industrial, commercial and domestic applications that can operate at essentially constant speed. Induction machines have simpler and more rugged structure, higher maintainability and economy than DC motors. They are also robust and immune to heavy loading.

In the mid 1980’s, a new advanced control technique, known as direct torque and flux control (DTFC or DTC) or direct self-control (DSC) [6] was introduced for voltage-fed PWM inverter drives. However, some research is still being done to adapt DTC to new motors and converters and also to reduce the torque ripple [1, 4] which is one of its main drawbacks. DTC, as the name indicates, is the direct control of torque and flux of a drive by the selection, through a look-up table, of the inverter voltage space vectors. The main advantage of DTC is its structure simplicity, since no coordinate transformations and no PWM generation are needed. However, torque and flux modulus values and the sector of the flux are needed. Not only it is a very simple and robust signal processing scheme but also a very quick and precise torque control response is achieved. Recently, three phase Matrix Converters (MC) have emerged to become a good alternative to the well known Voltage Source Inverter (VSI)[2,10] due to its advantages over voltage source inverters.

In this paper, a new control technique for matrix converters which generates, under unity input power factor conditions, the voltage vectors needed to implement the DTC for induction motors is investigated. Simulation results, which demonstrate the effectiveness of the proposed scheme, are presented. The results obtained states a good torque response and minimized torque ripples.

II. DIRECT TORQUE CONTROL THEORY

The principle of Direct Torque Control (DTC) is to directly select voltage vectors according to the difference between reference and actual value of torque and flux linkage. Torque and flux errors are compared in hysteresis comparators. Depending on the comparators, a voltage vector is selected from a table. Advantages [6] of the DTC are ; low complexity, no pulse width modulation is needed, no co-ordinate transformations needed. Instead one of the six VSI voltage vectors is applied during the whole sample period. All calculations are done in a stationary reference frame which does not involve the explicit knowledge of rotor position. Still, for a synchronous motor, rotor position must be known at start-up.

The DTC hence require low computational power when implemented digitally. The system possess good dynamic performance but shows quite poor
performance in steady-state since the crude voltage selection criteria give rise to high ripple levels in stator current, flux linkage and torque. Its simplicity makes it possible to execute every computational cycle in a short time period and use a high sample frequency. For every doubling in sample frequency, the ripple will approximately halve. The problem is that the power switches used in the inverter impose a limit for the maximum sample frequency.

In general, a scalar control technique called Volts/hertz control being among the simplest control methods. It is used where exact torque and flux control is not essential but where speed control is desirable, like when several motors are driven in parallel by a single inverter. Vector control is used where high performance torque and flux control is needed. Vector control can be implemented by using FOC or DTC technique. DTC technique is simple, robust and offer good dynamic performance. FOC technique gives the best performance in terms of ripples level but requires more processor power and are more complicated to implement.

III. DIRECT TORQUE CONTROL PRINCIPLE

In the direct torque control principles, it has been proven that the change of torque can be controlled by keeping the amplitude of the stator flux linkage constant and increasing the rotating speed of the stator flux linkage as fast as possible [6]. The amplitude and rotating speed of the stator flux linkage can be controlled by selecting the proper stator voltage vectors. Direct torque control (DTC) is one method used in variable frequency drives to control the torque (and thus finally the speed) of three-phase AC electric motors. This involves calculating an estimate of the motor’s magnetic flux and torque based on the measured voltage and current of the motor.

The torque expression can be expressed in the vector form as,

\[ T_e = \frac{3}{2} \left( \frac{P}{2} \right) \bar{\varphi}_s \times \bar{I}_s \]  

...(1)

where

\[ \bar{\varphi}_s = \varphi_{qs}^s - j \varphi_{ds}^s \]  

...(2)

and \[ \bar{I}_s = I_{qs}^s - j I_{ds}^s \]  

...(3)

In this equation, \( \bar{I}_s \) is to be replaced by the rotor flux \( \bar{\varphi} \). In the complex form, \( \bar{\varphi}_s \) and \( \bar{\varphi}_r \) can be expressed as the functions of currents as

\[ \bar{\varphi} = L_s \bar{I}_s + L_m \bar{I}_r \]  

...(4)

\[ \bar{\varphi} = L_r \bar{I}_r + L_m \bar{I}_s \]  

...(5)

Eliminating \( \bar{I}_r \) from equation \( \bar{\varphi}_s \), we get

\[ \bar{\varphi}_s = \frac{L_m}{L_r} \bar{\psi}_{br} + L_s' \bar{I}_s \]  

...(6)

where

\[ L_s' = L_s - L_r^2 \]  

...(7)

The corresponding equation of \( \bar{I}_s \) is

\[ \bar{I}_s = \frac{\bar{\varphi}_s}{L_s} - L_m \frac{\bar{\varphi}_r}{L_r L_s} \]  

...(8)

On substituting \( \bar{I}_s \) in \( \bar{T}_e \) and further simplification which yields,

\[ \bar{T}_e = \frac{3}{2} \left( \frac{P}{2} \right) \left( \frac{L_m}{L_r L_s} \right) \bar{\varphi}_r X \bar{\varphi}_s \]  

...(9)

that is, the magnitude of torque is,

\[ T_e = \frac{3}{2} \left( \frac{P}{2} \right) \left( \frac{L_m}{L_r L_s} \right) |\bar{\varphi}_r| |\bar{\varphi}_s| \sin \gamma \]  

...(10)

where \( \gamma \) is the angle between the fluxes.

If the rotor flux remains constant and the stator flux is changed incrementally by the stator voltage and the corresponding change of \( \gamma \) angle is \( \Delta \gamma \), the incremental torque expression \( \Delta T_e \) is given as

\[ \Delta T_e = \frac{3}{2} \left( \frac{P}{2} \right) \left( \frac{L_m}{L_r L_s} \right) |\bar{\varphi}_r| |\bar{\varphi}_s + \Delta \bar{\varphi}_s| \sin \Delta \gamma \]  

...(11)

IV. MATRIX CONVERTER

Nowadays Matrix Converters are being alternative for Voltage Source Inverters, which converts AC-AC supply without any intermediate dc link. The Matrix Converter is an advanced circuit topology capable of converting AC-AC, providing generation of load voltage with arbitrary amplitude and frequency, bidirectional power flow, sinusoidal input/output waveforms, and operation with unity input power factor. Furthermore,
since no inductive or capacitive elements are required, matrix converter allows a very compact design. A Matrix Converter is an AC-AC converter, with m x n bidirectional switches, which connects an m-phase voltage source to an n-phase load. From a practical point of view, the three-phase matrix converters having 3 x 3 switches connects a three phase voltage source to a three-phase load, e.g. a motor.

![Three Phase Matrix Converters](image)

In the MC shown in Fig.1, a switch \( S_{ij} = \begin{cases} i, & \text{if } i \in \{A, B, C\}, j \in \{a, b, c\} \text{ connects phase } i \text{ of the input to phase } j \text{ of the load.} \\
0, & \text{if } S_{ij} \text{ is open and } S_{ij} \text{ is closed.}
\end{cases} 
\)

The induction motor drive fed by matrix converter is superior to the induction motors fed by the conventional inverters because of the lack of bulky direct current (dc)-link capacitors with limited lifetime, bidirectional power flow capability, sinusoidal input/output currents and adjustable input power factor.

A mathematical model of the MC can be derived as follows:

Applying Kirchhoff’s voltage law, (12) can be easily obtained.

\[
\begin{bmatrix}
V_{an}(t) \\
V_{bn}(t) \\
V_{cn}(t)
\end{bmatrix}
= \begin{bmatrix}
S_{AN}(t) & S_{BA}(t) & S_{CA}(t) \\
S_{AB}(t) & S_{BB}(t) & S_{CB}(t) \\
S_{AC}(t) & S_{BC}(t) & S_{CA}(t)
\end{bmatrix}
\begin{bmatrix}
V_{AN}(t) \\
V_{BN}(t) \\
V_{CN}(t)
\end{bmatrix}
\tag{12}
\]

where the output voltages and output currents are being synthesized by the input voltages and input currents respectively.

V. SPACE VECTORS OF MATRIX CONVERTER

It can be noted that the three-phase MC allows any output phase to be connected to any input phase. Thus, 27 switching configuration are possible. Applying Clark’s transformation to the 27 possible switching configuration, it can be easily found that a MC can generate 18 active, voltage and current space vectors, three zero vectors and six rotating vectors [13]. Table 1 shows the 27 switching states of the three phase MC. The first 18 states \((\pm 1, \pm 2, \pm 3, \pm 4, \pm 5, \pm 6, \pm 7, \pm 8, \pm 9)\) corresponds to the active vectors. 0a, 0b and 0c are the corresponding zero vectors states. And the last six states (shadowed rows) are the rotating vectors. Representation of the output line-to-neutral voltages and input line currents space active vectors are shown in fig.2

![Space Vectors of Matrix Converters](image)

Fig.2 shows the space vectors for the Matrix Converters. The magnitude and direction of the active vectors depends upon the instantaneous line-to-neutral voltages and line currents as shown in Table 2. The direction of each of these vectors remains constant. On the contrary, the magnitude of the six rotating vectors remains constant and corresponds to the maximum value of the input line-to-neutral voltage vector and the input line current vector, while its direction depends on the angles of the line-to-neutral input voltage vector \(\alpha\) and the input line current vector. Fig.3 shows the vector rotations of Matrix Converters, which is used for the vector selection.
VI DTC USING MATRIX CONVERTERS

Matrix Converters generates a higher number of output voltage vectors with respect to a VSI, the introduction of a third variable, such as the average value of the sine of the displacement angle $\phi_i$ between the input line current vector and the input line to neutral voltage vector, can be used to keep under control of the input power factor. The flux, torque and the average value of sine $\phi_i$ estimations require the knowledge of voltages and currents at the input and output side of the MC. Assuming that $V_i$ is the output voltage vector selected by the basic DTC control algorithm, from table 1, it is obvious that one of the switching states $\pm 1, \pm 2$ or $\pm 3$ must be selected. Since the magnitude and direction of these voltage vectors depends on the input line-to-neutral voltage vector, only those having the same direction of $V_i$ and the maximum magnitude will be taken into consideration.
VII. SIMULATION RESULTS AND DISCUSSION

Fig. 4 shows the overall MATLAB/Simulink diagram for the Direct Torque Control of Matrix Converter fed Induction Motor drive.

Fig. 5 shows the variation of stator flux waveforms, where the stator flux is sinusoidal in nature, whose amplitude is about 0.7 wb/m2.

Fig. 6 shows the voltage sector waveforms of the Matrix Converter, in which the proper voltage sectors are being selected from the look up table in order to drive the Induction motor.

Table 2. Switching table for DTC using Matrix Converters

<table>
<thead>
<tr>
<th></th>
<th>1</th>
<th>2</th>
<th>3</th>
<th>4</th>
<th>5</th>
<th>6</th>
</tr>
</thead>
<tbody>
<tr>
<td>ϕ</td>
<td>+1</td>
<td>-1</td>
<td>+1</td>
<td>-1</td>
<td>+1</td>
<td>-1</td>
</tr>
<tr>
<td></td>
<td>-3</td>
<td>1</td>
<td>2</td>
<td>-3</td>
<td>1</td>
<td>2</td>
</tr>
<tr>
<td>V1</td>
<td>3</td>
<td>-7</td>
<td>8</td>
<td>9</td>
<td>-8</td>
<td>8</td>
</tr>
<tr>
<td>V2</td>
<td>-6</td>
<td>4</td>
<td>5</td>
<td>-6</td>
<td>5</td>
<td>-6</td>
</tr>
<tr>
<td>V3</td>
<td>3</td>
<td>-1</td>
<td>2</td>
<td>3</td>
<td>-1</td>
<td>2</td>
</tr>
<tr>
<td>V4</td>
<td>-9</td>
<td>7</td>
<td>8</td>
<td>-9</td>
<td>8</td>
<td>-9</td>
</tr>
<tr>
<td>V5</td>
<td>6</td>
<td>-4</td>
<td>5</td>
<td>6</td>
<td>-4</td>
<td>5</td>
</tr>
<tr>
<td>V6</td>
<td>9</td>
<td>-7</td>
<td>8</td>
<td>9</td>
<td>-7</td>
<td>8</td>
</tr>
</tbody>
</table>
Fig. 7. Stator Voltage of Induction Motor

Fig. 7 shows the stator voltage waveforms which is used to drive the induction motor, whose amplitude is about 300V, which is sufficient to drive the motor.

Fig. 8. Stator Current of Induction Motor

Fig. 8 shows the variation of stator current with respect to time where the initial starting current is about 25 A and then gradually decreases to about 15 A, which avoids the heating of windings.

Fig. 9. Rotor Flux of Induction Motor

Fig. 9 shows the variation of rotor flux with respect to time, in which the rotor flux amplitude is about 0.7 wb/m² and it is also sinusoidal as that of stator fluxes.

Fig. 10. Speed of Induction Motor

Fig. 10 shows the variation of speed with respect to time. At the time of starting, the speed gradually increases and then maintains the constant speed of about 1900rpm.

Fig. 11. Torque of Induction Motor

Fig. 11 shows the variation of torque with respect to time. When the speed starts increasing, the torque gradually decreases to about 40 Nm.

VII CONCLUSION

In this paper, the matrix converter fed induction motor drive implemented by DTC is simulated. Based on the principles of the matrix converter and direct torque control, combined control strategy is presented, which can implement the real-time space vector modulation of converter and induction motor stator flux linkage oriented direct torque control. The equations show that the change of torque can be controlled by keeping the amplitude of the stator flux linkage constant and increasing the rotating speed of the stator flux linkage as fast as possible. The amplitude and rotating speed of the stator flux linkage can be controlled by selecting the proper stator voltage vectors. The torque and flux linkage reference are kept constant at the same level in order to reduce the torque ripples. The simulated results shows that induction motor drive fed by the matrix converters using direct torque control technique has reduced torque ripples and faster torque response.
VIII REFERENCES


M. Suresh was born on 1988 in Tamilnadu, India. He received his B.E degree in Electrical and Electronics Engineering from Muthayammal Engineering College, Rasipuram, Affiliated to Anna University, Chennai in the year 2005. Currently, he is doing his Master's degree in Power Electronics and Drives in Bannari Amman Institute of Technology, Sathyamangalam, Affiliated to Anna University, Coimbatore. His area of interests includes Digital systems and Design, Sensorless control of Induction motor Drives.

J. Karpagam, She completed her Masters degree in the specialization of Power Electronics and Drives from Anna University Chennai, India. Presently she is working as Assistant Professor in the department of Electrical and Electronics Engineering at Bannari Amman Institute of Technology Tamilnadu, India and she is in this profession since 1999. Currently she is pursuing her Ph.D in the area of multilevel converters applications to AC Drives under Anna University Coimbatore. Her area of interest includes Power Electronics, DSP applications to Power Electronics and Wind Power Generation, Solid State Drives, Power Converters, AC & DC drives. So far she has published around ten papers in national level and international conferences.